

# DESIGN NOTE 

## Single IC, Five Output Switching Power Supply System for Portable Electronics - Design Note 150

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The drive for higher performance portable electronic systems, concomitant with the need to remain compatible with existing interfacing standards and hardware, has caused the number of system power supply voltages to proliferate. This Design Note describes how to generate five separate output voltages using the LTC ${ }^{\circledR}$ 1538-AUX, a dual synchronous switching regulator controller with internal circuitry adaptable to many additional configurations. The five output voltages chosen in this example are:

- $5 \mathrm{~V} \pm 4 \% / 25 \mathrm{~mA}$ linear regulator, which remains active regardless of the state of the switching regulator controllers
- $5 \mathrm{~V} \pm 2 \% / 3 \mathrm{~A}$ synchronous switching buck regulator
- $3.3 \mathrm{~V} \pm 2 \% / 6 \mathrm{~A}$ synchronous switching buck regulator
- $2.9 \mathrm{~V} \pm 5 \% / 3 \mathrm{~A}$ peak low dropout linear regulator deriving power from the 3.3 V output
- $12 \mathrm{~V} \pm 5 \% / 200 \mathrm{~mA}$ synchronously rectified flyback output deriving power from the 5 V output
The $100 \mu \mathrm{~A}$ quiescent current, 5 V standby linear regulator can efficiently power "wake-up" circuitry in portable systems and can deliver 25 mA .


Figure 1. 2.9V Output Transient Load Step Performance

The 5 V and 3.3 V switching regulators, independently activated by the two RUN/SS pins, provide very efficient, constant frequency operation. This is accomplished by using a synchronous-buck architecture at high currents and switching over to Burst Mode ${ }^{\text {TM }}$ operation below approximately $10 \%$ to $15 \%$ of maximum current, as determined by the current sensing resistor for each controller.
The 5V output from the first controller, or any external voltage between 4.8 V and 10 V , can be tied to the ExtV $\mathrm{V}_{\mathrm{C}}$ pin. The power and voltage from this external source will supplant the internal 5 V linear regulator if the applied voltage is greater than 4.8 V . This technique improves efficiency by eliminating the power dissipated by the IC due to the current drawn through the internal linear regulator and the $\left(\mathrm{V}_{\mathrm{IN}}-5 \mathrm{~V}\right)$ voltage drop.
The 2.9V linear regulator draws power from the 3.3 V output and performs the function of a wide bandwidth, low dropout linear regulator having dynamic performance as illustrated in Figure 1.
The regulator's dominant pole is set by the output capacitance and the load resistance. The loop is stable, provided there is some $\operatorname{ESR}(0.02 \Omega$ to $0.1 \Omega)$ in the output capacitor in order to generate phase lead prior to the unity-gain crossover frequency. The AVX-TPS tantalum capacitor used here, or a Sanyo OS-CON type capacitor, has a complex impedance characteristic, providing close to zero phase shift at the unity-gain cross frequency of the amplifier.
The $12 \mathrm{~V} / 200 \mathrm{~mA}$ output uses a synchronously driven, tightly coupled secondary winding on the first controller's primary winding to generate a highly efficient output with a $\pm 5 \%$ tolerance under all load and line conditions. The output voltage is fed back to the SFB1 input pin and compared with the internal 1.19 V reference. This feedback
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forces the primary controller into a "forced synchronous" mode, regulating the 12 V output regardless of the loading of the primary 5 V regulator. Tighter regulation and less ripple can be attained with a slightly higher turns ratio and an additional linear regulator.

Figure 2 shows the overall efficiency for the circuit shown in Figure 3. The $\mathrm{I}_{2.9}=$ Proportional curve documents efficiency when all outputs are loaded proportionally to their peak design loads of: 5 V at $3 \mathrm{~A}, 3.3 \mathrm{~V}$ at $3 \mathrm{~A}, 2.9 \mathrm{~V}$ at 3 A and 12 V at 200 mA . As an example, overall efficiency drops to $90 \%$ when the 2.9 V output is loaded with 1 A ( $33 \%$ of the designed maximum load). For this same efficiency point, the three other outputs are loaded as follows: 5 V at $1 \mathrm{~A}, 3.3 \mathrm{~V}$ at 1 A and 12 V at 67 mA . For the peak load of 36 W , as specified above, efficiency drops to $85 \%$.
The $\mathrm{I}_{2.9}=0$ curve shows efficiency when all outputs are loaded proportional to their peak loads except the 2.9 V
output which is not loaded. This curve documents a $100 \%$ design load of 27.3 W . The circuit's efficiency peaks for this curve at $95 \%$ when the 5 V and 3.3 V outputs are loaded with 1 A and the 12 V output is loaded with 67 mA .


Figure 2. Efficiency vs Percent of the Designed Maximum Load


Figure 3. Schematic Diagram of System

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